General Description

The MIC5205 is an efficient linear voltage regulator with ultra-low-noise output, very low dropout voltage (typically 17mV at light loads and 165mV at 150mA), and very low ground current (600µA at 100mA output). The MIC5205 offers better than 1% initial accuracy.

Designed especially for hand-held, battery-powered devices, the MIC5205 includes a CMOS or TTL compatible enable/shutdown control input. When shutdown, power consumption drops nearly to zero. Regulator ground current increases only slightly in dropout, further prolonging battery life.

Key MIC5205 features include a reference bypass pin to improve its already excellent low-noise performance, reversed-battery protection, current limiting, and overtemperature shutdown.

The MIC5205 is available in fixed and adjustable output voltage versions in a small SOT-23-5 package.

For low-dropout regulators that are stable with ceramic output capacitors, see the μCap MIC5245/6/7 family.

Features

- Ultra-low-noise output
- High output voltage accuracy
- Guaranteed 150mA output
- Low quiescent current
- Low dropout voltage
- Extremely tight load and line regulation
- Very low temperature coefficient
- Current and thermal limiting
- Reverse-battery protection
- "Zero" off-mode current
- Logic-controlled electronic enable

Applications

- Cellular telephones
- Laptop, notebook, and palmtop computers
- Battery-powered equipment
- PCMCIA Vcc and Vpp regulation/switching
- Consumer/personal electronics
- SMPS post-regulator/dc-to-dc modules
- High-efficiency linear power supplies

Ordering Information

<table>
<thead>
<tr>
<th>Part Number</th>
<th>Marking</th>
<th>Voltage</th>
<th>Accuracy</th>
<th>Junction Temp. Range*</th>
<th>Package</th>
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<tbody>
<tr>
<td>MIC5205BM5</td>
<td>LBAA</td>
<td>Adj</td>
<td>1%</td>
<td>–40°C to +125°C</td>
<td>SOT-23-5</td>
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<tr>
<td>MIC5205-2.5BM5</td>
<td>LB25</td>
<td>2.5V</td>
<td>1%</td>
<td>–40°C to +125°C</td>
<td>SOT-23-5</td>
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<tr>
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<td>–40°C to +125°C</td>
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<tr>
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<td>LB28</td>
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<td>–40°C to +125°C</td>
<td>SOT-23-5</td>
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<td>–40°C to +125°C</td>
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<td>LB29</td>
<td>2.9V</td>
<td>1%</td>
<td>–40°C to +125°C</td>
<td>SOT-23-5</td>
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<tr>
<td>MIC5205-3.0BM5</td>
<td>LB30</td>
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<td>1%</td>
<td>–40°C to +125°C</td>
<td>SOT-23-5</td>
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<tr>
<td>MIC5205-3.8BM5</td>
<td>LB38</td>
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<td>1%</td>
<td>–40°C to +125°C</td>
<td>SOT-23-5</td>
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<tr>
<td>MIC5205-4.0BM5</td>
<td>LB40</td>
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<td>1%</td>
<td>–40°C to +125°C</td>
<td>SOT-23-5</td>
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<tr>
<td>MIC5205-5.0BM5</td>
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<td>5.0V</td>
<td>1%</td>
<td>–40°C to +125°C</td>
<td>SOT-23-5</td>
</tr>
</tbody>
</table>

Other voltages available. Contact Micrel for details.

Typical Application

Ultra-Low-Noise Regulator Application
Pin Configuration

**Absolute Maximum Ratings (Note 1)**

- Supply Input Voltage ($V_{IN}$) \(-20\text{V} \text{ to } +20\text{V}\)
- Enable Input Voltage ($V_{EN}$) \(-20\text{V} \text{ to } +20\text{V}\)
- Power Dissipation ($P_D$) \(\text{Internally Limited, Note 3}\)
- Lead Temperature (soldering, 5 sec.) \(260\degree\text{C}\)
- Junction Temperature ($T_J$) \(-40\degree\text{C} \text{ to } +125\degree\text{C}\)
- Storage Temperature ($T_S$) \(-65\degree\text{C} \text{ to } +150\degree\text{C}\)

**Operating Ratings (Note 2)**

- Input Voltage ($V_{IN}$) \(+2.5\text{V} \text{ to } +16\text{V}\)
- Enable Input Voltage ($V_{EN}$) \(0\text{V} \text{ to } V_{IN}\)
- Junction Temperature ($T_J$) \(-40\degree\text{C} \text{ to } +125\degree\text{C}\)
- Thermal Resistance, SOT-23-5 ($\theta_{JA}$) \(\text{Note 3}\)
# Electrical Characteristics

$V_{IN} = V_{OUT} + 1V; I_L = 100\mu A; C_L = 1.0\mu F; V_{EN} \geq 2.0V; T_J = 25^\circ C$, **bold** values indicate $-40^\circ C \leq T_J \leq +125^\circ C$; unless noted.

<table>
<thead>
<tr>
<th>Symbol</th>
<th>Parameter</th>
<th>Conditions</th>
<th>Min</th>
<th>Typical</th>
<th>Max</th>
<th>Units</th>
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<tbody>
<tr>
<td>$V_O$</td>
<td>Output Voltage Accuracy</td>
<td>variation from specified $V_{OUT}$</td>
<td>$-1$</td>
<td>$1$</td>
<td></td>
<td>%</td>
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<tr>
<td>$\Delta V_O/\Delta T$</td>
<td>Output Voltage Temperature Coefficient</td>
<td>Note 4</td>
<td>$40$</td>
<td></td>
<td></td>
<td>ppm/°C</td>
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<td>$\Delta V_O/V_O$</td>
<td>Line Regulation</td>
<td>$V_{IN} = V_{OUT} + 1V$ to $16V$</td>
<td>$0.004$</td>
<td>$0.05$</td>
<td></td>
<td>% / V</td>
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<tr>
<td>$\Delta V_O/V_O$</td>
<td>Load Regulation</td>
<td>$I_L = 0.1mA$ to $150mA$, Note 5</td>
<td>$0.02$</td>
<td>$0.5$</td>
<td></td>
<td>%</td>
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<td>$V_{IN} - V_O$</td>
<td>Dropout Voltage, Note 6</td>
<td>$I_L = 100\mu A$</td>
<td>$10$</td>
<td>$50$</td>
<td>mV</td>
<td></td>
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<tr>
<td></td>
<td></td>
<td>$I_L = 50mA$</td>
<td>$110$</td>
<td>$70$</td>
<td>mV</td>
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<td></td>
<td></td>
<td>$I_L = 100mA$</td>
<td>$140$</td>
<td>$230$</td>
<td>mV</td>
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<tr>
<td></td>
<td></td>
<td>$I_L = 150mA$</td>
<td>$165$</td>
<td>$300$</td>
<td>mV</td>
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<td>$I_{GND}$</td>
<td>Quiescent Current</td>
<td>$V_{EN} \leq 0.4V$ (shutdown)</td>
<td>$0.01$</td>
<td>$1$</td>
<td></td>
<td>μA</td>
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<tr>
<td></td>
<td></td>
<td>$V_{EN} \leq 0.18V$ (shutdown)</td>
<td></td>
<td>$5$</td>
<td>μA</td>
<td></td>
</tr>
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<td>$I_{GND}$</td>
<td>Ground Pin Current, Note 7</td>
<td>$V_{EN} \geq 2.0V$, $I_L = 100\mu A$</td>
<td>$80$</td>
<td>$125$</td>
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<td>μA</td>
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<td></td>
<td></td>
<td>$I_L = 50mA$</td>
<td>$350$</td>
<td>$150$</td>
<td>μA</td>
<td></td>
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<tr>
<td></td>
<td></td>
<td>$I_L = 100mA$</td>
<td>$600$</td>
<td>$800$</td>
<td>μA</td>
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<td>$I_L = 150mA$</td>
<td>$1300$</td>
<td>$1500$</td>
<td>μA</td>
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<td>$I_L = 200mA$</td>
<td>$2500$</td>
<td></td>
<td>μA</td>
<td></td>
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<td>PSRR</td>
<td>Ripple Rejection</td>
<td>frequency = $100Hz$, $I_L = 100\mu A$</td>
<td>$75$</td>
<td></td>
<td></td>
<td>dB</td>
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<tr>
<td>$I_{LIMIT}$</td>
<td>Current Limit</td>
<td>$V_{OUT} = 0V$</td>
<td>$320$</td>
<td>$500$</td>
<td></td>
<td>mA</td>
</tr>
<tr>
<td>$\Delta V_O/\Delta P_D$</td>
<td>Thermal Regulation</td>
<td>Note 8</td>
<td>$0.05$</td>
<td></td>
<td></td>
<td>%/W</td>
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<tr>
<td>$e_{no}$</td>
<td>Output Noise</td>
<td>$I_L = 50mA$, $C_L = 2.2\mu F$, $470pF$ from BYP to GND</td>
<td>$260$</td>
<td></td>
<td></td>
<td>nV/√Hz</td>
</tr>
</tbody>
</table>

**ENABLE Input**

| $V_{IL}$ | Enable Input Logic-Low Voltage | regulator shutdown | $0.4$ | $0.18$ | | V |
| $V_{IH}$ | Enable Input Logic-High Voltage | regulator enabled | $2.0$ | | | V |
| $I_{IL}$ | Enable Input Current | $V_{IL} \leq 0.4V$ | $0.01$ | | | μA |
| | | $V_{IL} \leq 0.18V$ | $2$ | | | μA |
| | | $V_{IH} \geq 2.0V$ | $5$ | | | μA |

**Note 1.** Exceeding the absolute maximum rating may damage the device.

**Note 2.** The device is not guaranteed to function outside its operating rating.

**Note 3.** The maximum allowable power dissipation at any $T_A$ (ambient temperature) is $P_{D(max)} = (T_{J(max)} - T_A) / \theta_JA$. Exceeding the maximum allowable power dissipation will result in excessive die temperature, and the regulator will go into thermal shutdown. The $\theta_JA$ of the MIC5205-xxBM5 (all versions) is $220^\circ C/W$ mounted on a PC board (see “Thermal Considerations” section for further details).

**Note 4.** Output voltage temperature coefficient is defined as the worst case voltage change divided by the total temperature range.

**Note 5.** Regulation is measured at constant junction temperature using low duty cycle pulse testing. Parts are tested for load regulation in the load range from $0.1mA$ to $150mA$. Changes in output voltage due to heating effects are covered by the thermal regulation specification.

**Note 6.** Dropout Voltage is defined as the input to output differential at which the output voltage drops $2\%$ below its nominal value measured at $1V$ differential.

**Note 7.** Ground pin current is the regulator quiescent current plus pass transistor base current. The total current drawn from the supply is the sum of the load current plus the ground pin current.

**Note 8.** Thermal regulation is defined as the change in output voltage at a time “$t$” after a change in power dissipation is applied, excluding load or line regulation effects. Specifications are for a $150mA$ load pulse at $V_{IN} = 16V$ for $t = 10ms$. 

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June 2000
Typical Characteristics

- **Power Supply Rejection Ratio**
  - **Power Supply** Rejection Ratio
  - **FREQUENCY (Hz)**
  - **PSRR (dB)**
  - **V IN** = 6V
  - **V OUT** = 5V
  - **I OUT** = 100mA
  - **C OUT** = 1µF

- **Power Supply Rejection Ratio**
  - **Power Supply** Rejection Ratio
  - **FREQUENCY (Hz)**
  - **PSRR (dB)**
  - **V IN** = 6V
  - **V OUT** = 5V
  - **I OUT** = 1mA
  - **C OUT** = 1µF
  - **C BYP** = 0.01µF

- **Power Supply Ripple Rejection**
  - **Power Supply Ripple Rejection**
  - **VOLTAGE DROP (V)**
  - **RIPPLE REJECTION (dB)**
  - **V IN** = 6V
  - **V OUT** = 5V
  - **I OUT** = 100mA
  - **I OUT** = 10mA
  - **I OUT** = 1mA
  - **C OUT** = 2.2µF
  - **C BYP** = 0.01µF

- **Turn-On Time vs. Bypass Capacitance**
  - **Turn-On Time**
  - **CAPACITANCE (pF)**
  - **V IN** = 6V
  - **V OUT** = 5V
  - **I OUT** = 100mA
  - **I OUT** = 10mA
  - **I OUT** = 1mA
  - **C OUT** = 2.2µF
  - **C BYP** = 0.01µF

- **Dropout Voltage vs. Output Current**
  - **Dropout Voltage**
  - **OUTPUT CURRENT (mA)**
  - **V IN** = 6V
  - **V OUT** = 5V
  - **I OUT** = 100mA
  - **I OUT** = 10mA
  - **I OUT** = 1mA
  - **C OUT** = 2.2µF
  - **C BYP** = 0.01µF

- **Output Voltage vs. Temperature**
  - **VOLTAGE DROP (V)**
  - **DROPOUT VOLTAGE (mV)**
  - **V IN** = 6V
  - **V OUT** = 5V
  - **I OUT** = 100mA
  - **I OUT** = 10mA
  - **I OUT** = 1mA
  - **C OUT** = 2.2µF
  - **C BYP** = 0.01µF

- **Dropout Voltage vs. Output Current**
  - **Dropout Voltage**
  - **OUTPUT CURRENT (mA)**
  - **V IN** = 6V
  - **V OUT** = 5V
  - **I OUT** = 100mA
  - **I OUT** = 10mA
  - **I OUT** = 1mA
  - **C OUT** = 2.2µF
  - **C BYP** = 0.01µF

- **Turn-On Time vs. Bypass Capacitance**
  - **Turn-On Time**
  - **CAPACITANCE (pF)**
  - **V IN** = 6V
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- **Dropout Voltage vs. Output Current**
  - **Dropout Voltage**
  - **OUTPUT CURRENT (mA)**
  - **V IN** = 6V
  - **V OUT** = 5V
  - **I OUT** = 100mA
  - **I OUT** = 10mA
  - **I OUT** = 1mA
  - **C OUT** = 2.2µF
  - **C BYP** = 0.01µF
Typical Characteristics

Noise Performance

<table>
<thead>
<tr>
<th>Noise Performance</th>
<th>Noise Performance</th>
<th>Noise Performance</th>
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<tbody>
<tr>
<td>NOISE (µV/√Hz)</td>
<td>NOISE (µV/√Hz)</td>
<td>NOISE (µV/√Hz)</td>
</tr>
<tr>
<td>FREQUENCY (Hz)</td>
<td>FREQUENCY (Hz)</td>
<td>FREQUENCY (Hz)</td>
</tr>
</tbody>
</table>

- 10mA, C<sub>OUT</sub> = 1µF
- C<sub>BYP</sub> = 10nF

- V<sub>OUT</sub> = 5V
- C<sub>OUT</sub> = 1µF
- C<sub>BYP</sub> = 10nF

- 10mA, C<sub>OUT</sub> = 1µF
- C<sub>BYP</sub> = 100pF

- V<sub>OUT</sub> = 5V
- C<sub>OUT</sub> = 1µF
- C<sub>BYP</sub> = 1nF

- 10mA, C<sub>OUT</sub> = 1µF
- C<sub>BYP</sub> = 10nF

- V<sub>OUT</sub> = 5V
- C<sub>OUT</sub> = 1µF
- C<sub>BYP</sub> = 10nF

<table>
<thead>
<tr>
<th>Noise Performance</th>
<th>Noise Performance</th>
<th>Noise Performance</th>
</tr>
</thead>
<tbody>
<tr>
<td>NOISE (µV/√Hz)</td>
<td>NOISE (µV/√Hz)</td>
<td>NOISE (µV/√Hz)</td>
</tr>
<tr>
<td>FREQUENCY (Hz)</td>
<td>FREQUENCY (Hz)</td>
<td>FREQUENCY (Hz)</td>
</tr>
</tbody>
</table>

- 10mA, C<sub>OUT</sub> = 1µF
- C<sub>BYP</sub> = 10nF

- V<sub>OUT</sub> = 5V
- C<sub>OUT</sub> = 1µF
- C<sub>BYP</sub> = 10nF

- 10mA, C<sub>OUT</sub> = 1µF
- C<sub>BYP</sub> = 100pF

- V<sub>OUT</sub> = 5V
- C<sub>OUT</sub> = 1µF
- C<sub>BYP</sub> = 1nF

- 10mA, C<sub>OUT</sub> = 1µF
- C<sub>BYP</sub> = 10nF

- V<sub>OUT</sub> = 5V
- C<sub>OUT</sub> = 1µF
- C<sub>BYP</sub> = 10nF
Block Diagrams

**Ultra-Low-Noise Fixed Regulator**

**Ultra-Low-Noise Adjustable Regulator**

V<sub>OUT</sub> = V<sub>REF</sub> (1 + R2/R1)
Applications Information

Enable/Shutdown

Forcing EN (enable/shutdown) high (> 2V) enables the regulator. EN is compatible with CMOS logic gates. If the enable/shutdown feature is not required, connect EN (pin 3) to IN (supply input, pin 1). See Figure 1.

Input Capacitor

A 1µF capacitor should be placed from IN to GND if there is more than 10 inches of wire between the input and the ac filter capacitor or if a battery is used as the input.

Reference Bypass Capacitor

BYP (reference bypass) is connected to the internal voltage reference. A 470pF capacitor (C\textsubscript{BYP}) connected from BYP to GND quiets this reference, providing a significant reduction in output noise. C\textsubscript{BYP} reduces the regulator phase margin; when using C\textsubscript{BYP}, output capacitors of 2.2µF or greater are generally required to maintain stability.

The start-up speed of the MIC5205 is inversely proportional to the size of the reference bypass capacitor. Applications requiring a slow ramp-up of output voltage should consider omitting C\textsubscript{BYP}. Likewise, if rapid turn-on is necessary, consider omitting C\textsubscript{BYP}.

If output noise is not a major concern, omit C\textsubscript{BYP} and leave BYP open.

Output Capacitor

An output capacitor is required between OUT and GND to prevent oscillation. The minimum size of the output capacitor is dependent upon whether a reference bypass capacitor is used. 1.0µF minimum is recommended when C\textsubscript{BYP} is not used (see Figure 2). 2.2µF minimum is recommended when C\textsubscript{BYP} is 470pF (see Figure 1). Larger values improve the regulator’s transient response. The output capacitor value may be increased without limit.

The output capacitor should have an ESR (effective series resistance) of about 5Ω or less and a resonant frequency above 1MHz. Ultra-low-ESR capacitors can cause a low amplitude oscillation on the output and/or underdamped transient response. Most tantalum or aluminum electrolytic capacitors are adequate; film types will work, but are more expensive. Since many aluminum electrolytics have electrolytes that freeze at about –30°C, solid tantalums are recommended for operation below –25°C.

At lower values of output current, less output capacitance is required for output stability. The capacitor can be reduced to 0.47µF for current below 10mA or 0.33µF for currents below 1mA.

No-Load Stability

The MIC5205 will remain stable and in regulation with no load (other than the internal voltage divider) unlike many other voltage regulators. This is especially important in CMOS RAM keep-alive applications.

Thermal Considerations

The MIC5205 is designed to provide 150mA of continuous current in a very small package. Maximum power dissipation can be calculated based on the output current and the voltage drop across the part. To determine the maximum power dissipation of the package, use the junction-to-ambient thermal resistance of the device and the following basic equation:

$$P_{D(max)} = \frac{(T_{J(max)} - T_A)}{\theta_{JA}}$$

$T_{J(max)}$ is the maximum junction temperature of the die, 125°C, and $T_A$ is the ambient operating temperature. $\theta_{JA}$ is layout dependent; Table 1 shows examples of junction-to-ambient thermal resistance for the MIC5205.

<table>
<thead>
<tr>
<th>Package</th>
<th>θ\textsubscript{JA} Recommended Minimum Footprint</th>
<th>θ\textsubscript{JA} 1&quot; Square Copper Clad</th>
<th>θ\textsubscript{JC}</th>
</tr>
</thead>
<tbody>
<tr>
<td>SOT-23-5 (M5)</td>
<td>220°C/W</td>
<td>170°C/W</td>
<td>130°C/W</td>
</tr>
</tbody>
</table>

Table 1. SOT-23-5 Thermal Resistance

The actual power dissipation of the regulator circuit can be determined using the equation:

$$P_D = (V_{IN} - V_{OUT}) I_{OUT} + V_{IN} I_{GND}$$

Substituting $P_{D(max)}$ for $P_D$ and solving for the operating conditions that are critical to the application will give the maximum operating conditions for the regulator circuit. For example, when operating the MIC5205-3.3BM5 at room temperature with a minimum footprint layout, the maximum input voltage for a set output current can be determined as follows:

$$P_{D(max)} = \frac{(125°C - 25°C)}{220°C/W}$$

Therefore, a 3.3V application at 150mA of output current can accept a maximum input voltage of 6.2V in a SOT-23-5 package. For a full discussion of heat sinking and thermal effects on voltage regulators, refer to the Regulator Thermals section of Micrel's Designing with Low-Dropout Voltage Regulators handbook.
Fixed Regulator Applications

Figure 1. Ultra-Low-Noise Fixed Voltage Application
Figure 1 includes a 470pF capacitor for low-noise operation and shows EN (pin 3) connected to IN (pin 1) for an application where enable/shutdown is not required. \( C_{OUT} = 2.2\mu F \) minimum.

Figure 2. Low-Noise Fixed Voltage Application
Figure 2 is an example of a low-noise configuration where \( C_{BYP} \) is not required. \( C_{OUT} = 1\mu F \) minimum.

Adjustable Regulator Applications
The MIC5205BM5 can be adjusted to a specific output voltage by using two external resistors (Figure 3). The resistors set the output voltage based on the following equation:

\[
V_{OUT} = 1.242V \times \left( \frac{R2}{R1} + 1 \right)
\]

This equation is correct due to the configuration of the bandgap reference. The bandgap voltage is relative to the output, as seen in the block diagram. Traditional regulators normally have the reference voltage relative to ground and have a different \( V_{OUT} \) equation.

Resistor values are not critical because ADJ (adjust) has a high input impedance, but for best results use resistors of 470k\( \Omega \) or less. A capacitor from ADJ to ground provides greatly improved noise performance.

Figure 3. Ultra-Low-Noise Adjustable Voltage Application
Figure 3 includes the optional 470pF noise bypass capacitor from ADJ to GND to reduce output noise.

Dual-Supply Operation
When used in dual supply systems where the regulator load is returned to a negative supply, the output voltage must be diode clamped to ground.
Package Information

DIMENSIONS:

MM (INCH)

1.90 (0.075) REF
0.95 (0.037) REF
1.75 (0.069)
1.50 (0.059)
3.02 (0.119)
2.80 (0.110)
0.50 (0.020)
0.35 (0.014)
0.15 (0.006)
0.00 (0.000)

SOT-23-5 (M5)

DIMENSIONS:

MM (INCH)

0.20 (0.008)
0.09 (0.004)
0.60 (0.024)
0.10 (0.004)